

SWR and BANDWIDTH OF SERIES / PARALLEL R L C

1- Series R L C circuit - Calculations and Simulations

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$$Z = j \cdot \omega \cdot L + \frac{1}{j \cdot \omega \cdot C} + R$$

Impedance of a series R L C

Eq 1

$$Q = \frac{\omega \cdot L}{R} \quad \text{Then :} \quad R = \frac{\omega \cdot L}{Q}$$

Q factors (unloaded !)

Eq 2

$$Q = \frac{1}{\omega \omega_0 \cdot C \cdot R} \quad \text{Then :} \quad R = \frac{1}{\omega \omega_0 \cdot C \cdot Q}$$

ω_0 is the resonant frequency

Eq 3

The impedance in Eq 1 is normalized

$$\frac{Z}{R} = j \cdot \omega \cdot \frac{L}{R} + \frac{1}{j \cdot \omega \cdot C \cdot R} + 1$$

Eq 4

$$R = \frac{\omega \omega_0 \cdot L}{Q} = \frac{1}{\omega \omega_0 \cdot C \cdot Q}$$

Eq 5

Substitute Eq 5 into Eq 4:

$$\frac{Z}{R} = j \cdot \omega \cdot \frac{L \cdot Q}{\omega \omega_0 \cdot L} + \frac{\omega \omega_0 \cdot C \cdot Q}{j \cdot \omega \cdot C} + 1$$

Eq 6

$$\frac{Z}{R} = j \cdot \omega \cdot \frac{Q}{\omega \omega_0} + \frac{\omega \omega_0 \cdot Q}{j \cdot \omega} + 1$$

After simplification

Eq 7

$$\frac{Z}{R} = j \cdot Q \cdot \left(\frac{\omega}{\omega \omega_0} - \frac{\omega \omega_0}{\omega} \right) + 1$$

Note that at resonance $\omega = \omega_0$
and $Z / R = 1$

Eq 8

At ω_H and ω_L the half power frequencies, the magnitude of Z/R equals $1 + j$
Therefore

$$\left| Q \cdot \left(\frac{\omega_H}{\omega \omega_0} - \frac{\omega \omega_0}{\omega_H} \right) \right| = \left| Q \cdot \left(\frac{\omega_L}{\omega \omega_0} - \frac{\omega \omega_0}{\omega_L} \right) \right| = 1$$

Eq 9

Then

$$\frac{\omega_H}{\omega \omega_0} - \frac{\omega \omega_0}{\omega_H} = \frac{\omega \omega_0}{\omega_L} - \frac{\omega_L}{\omega \omega_0}$$

Eq 10

Rearranging the terms:

$$\frac{\omega_H}{\omega \omega_0} + \frac{\omega_L}{\omega \omega_0} = \frac{\omega \omega_0}{\omega_L} + \frac{\omega \omega_0}{\omega_H}$$

Eq 11

$$\frac{\omega_H + \omega_L}{\omega\omega} = \frac{\omega\omega \cdot (\omega_H + \omega_L)}{\omega_H \cdot \omega_L} \quad \text{Eq 12}$$

Therefore

$$\omega_H \cdot \omega_L = \omega\omega^2 \quad \text{Eq 13}$$

$$\omega\omega = \sqrt{\omega_H \cdot \omega_L} \quad \text{Eq 14}$$

Rearranging the first term of eq 9:

$$Q \cdot \frac{(\omega_H)^2 - \omega\omega^2}{\omega\omega \cdot \omega_H} = 1 \quad \text{Eq 15}$$

From a combination of Eq 13 and Eq 15 :

$$Q \cdot \frac{(\omega_H)^2 - \omega_H \cdot \omega_L}{\omega\omega \cdot \omega_H} = 1 \quad \text{Eq 16}$$

$$Q \cdot \frac{\omega_H - \omega_L}{\omega\omega} = 1 \quad \text{Eq 17}$$

Then

$$\omega_H - \omega_L = \frac{\omega\omega}{Q} \quad \text{Eq 18}$$

Converting to frequency:

$$f_H - f_L = \frac{f_0}{Q} \quad \text{Eq 19}$$

f_0 is the resonant frequency
 f_H and f_L are the half power frequencies

The bandwidth BW in Hz is defined as:

$$BW = f_H - f_L \quad \text{Eq 20}$$

Then from Eq. 19

$$BW = \frac{f_0}{Q} \quad \text{And:} \quad Q = \frac{f_0}{BW} \quad \text{Eq 21a, 21b}$$

Eq 8 repeated

$$\frac{Z}{R} = j \cdot Q \cdot \left(\frac{\omega}{\omega\omega} - \frac{\omega\omega}{\omega} \right) + 1 \quad \text{Eq 8}$$

For the SERIES RLC circuit :

We have established that at $\omega = \omega_H$ and $\omega = \omega_L$ the **normalized** impedance is:

$$\frac{Z}{R} = 1 + j \quad \text{or} \quad \frac{Z}{R} = 1 - j$$

Note that at resonance $Z / R = 1$ from Eq. 8.
This is the minimum value.

The reflection coefficient at $\omega = \omega_H$ and $\omega = \omega_L$ is: **These are calculated with respect to the normalized Z_0 value of 1**

$$\text{rho} := \frac{(1 + j) - 1}{(1 + j) + 1} \quad \text{rho} = 0.2 + 0.4i \quad |\text{rho}| = 0.447 \quad \text{Reflection coefficient}$$

$$\text{rho} := \frac{(1 - j) - 1}{(1 - j) + 1} \quad \text{rho} = 0.2 - 0.4i \quad |\text{rho}| = 0.447 \quad \text{Reflection coefficient}$$

For the PARALLEL RLC circuit :

At $\omega = \omega_H$ and $\omega = \omega_L$ the **normalized** impedance Z/R is 1 in parallel with $+ / - j$:

Combining these impedances in parallel:

$$\frac{Z}{R} = \frac{1 \cdot j}{1 + j} = 0.5 + 0.5i$$

$$\frac{Z}{R} = \frac{1 \cdot (-j)}{1 - j} = 0.5 - 0.5i$$

The reflection coefficient at $\omega = \omega_H$ and $\omega = \omega_L$ is:

$$\text{rho} := \frac{(0.5 + 0.5i) - 1}{(0.5 + 0.5i) + 1} \quad \text{rho} = -0.2 + 0.4i \quad |\text{rho}| = 0.447 \quad \text{Reflection coefficient}$$

$$\text{rho} := \frac{(0.5 - 0.5i) - 1}{(0.5 - 0.5i) + 1} \quad \text{rho} = -0.2 - 0.4i \quad |\text{rho}| = 0.447 \quad \text{Reflection coefficient}$$

For BOTH series and parallel circuits, the reflection coefficients, return loss and SWR all have the same absolute value at the 3 dB points, where $\omega = \omega_H$ and $\omega = \omega_L$

$$\text{RL} := -20 \cdot \log(|\text{rho}|) \quad \text{RL} = 6.99 \quad \text{Return Loss}$$

$$\text{SWR}_t := \frac{1 + |\text{rho}|}{1 - |\text{rho}|} \quad \text{SWR}_t = 2.61803 \quad \text{SWR Reference Value}$$

Example of R L C SERIES CIRCUIT

$$f_r := 10 \text{ MHz} \quad L := 100 \text{ uH} \quad R := 50 \text{ ohms} \quad Z_o := R$$

$$C := \frac{10^6}{(2 \cdot \pi \cdot f_r)^2 \cdot L} \quad C = 2.533 \quad \text{C in pF is calculated to resonate at } f_r$$

$$X_L(f) := j \cdot 2 \cdot \pi \cdot f \cdot L \quad X_L(f_r) = 6.283i \times 10^3 \text{ Reactances}$$

$$X_C(f) := \frac{1}{j \cdot 2 \cdot \pi \cdot f \cdot C \cdot 10^{-6}} \quad X_C(f_r) = -6.283i \times 10^3$$

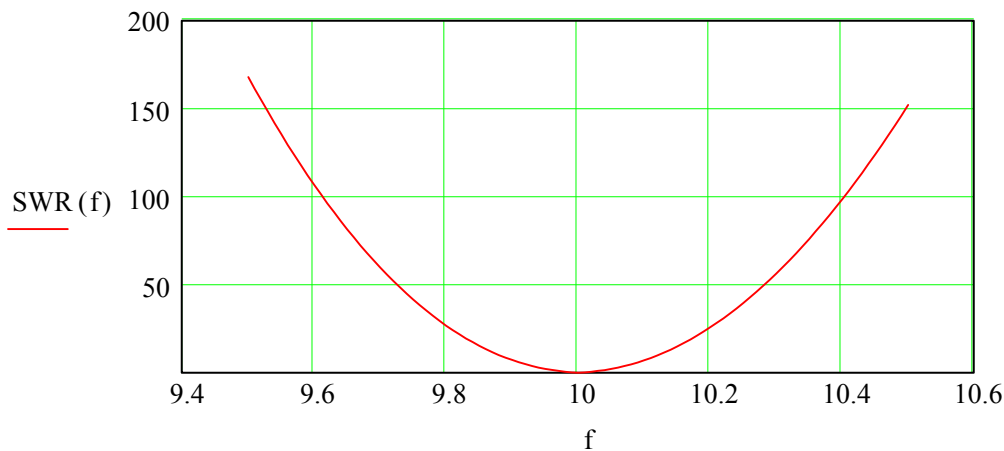
$$Z(f) := R + X_L(f) + X_C(f) \quad Z(f_r) = 50 \quad \text{At resonance}$$

$$\rho(f) := \left| \frac{Z(f) - Z_o}{Z(f) + Z_o} \right| \quad \rho(f_r) = 0 \quad \mathbf{\rho = 0}$$

$$\text{SWR}(f) := \frac{1 + \rho(f)}{1 - \rho(f)} \quad \text{SWR}(f_r) = 1$$

The minimum SWR should be 1:1 at f_r

$$f := 9.5, 9.51 \dots 10.5$$



Finding the two frequencies F1 and F2 that give SWR = 2.618

SWRt = 2.618 target SWR

Fa := 0.9998·fr start searching for f_low

Given

SWR(Fa) = SWRt

F1(SWRt) := Find(Fa) f_low := F1(SWRt) f_low = 9.96029

fr - f_low = 0.0397

Fb := 1.0002·fr start searching for f_hi

Given

SWR(Fb) = SWRt

F2(SWRt) := Find(Fb) f_hi := F2(SWRt) f_hi = 10.03987

f_hi - fr = 0.0399

Calculate the Relative Bandwidth Rel_BW (with respect to the resonant freq) at the target SWR

$$\text{Rel_BW}(\text{SWRt}) := \frac{F2(\text{SWRt}) - F1(\text{SWRt})}{fr} \qquad \text{Rel_BW}(\text{SWRt}) = 7.9578 \times 10^{-3}$$

We can calculate the Q at the target SWR = 2.618 (unloaded !)

$$Q := \frac{1}{\text{Rel_BW}(\text{SWRt})} \qquad Q = 125.664$$

Q from the circuit values, we get the same Q value: (unloaded !)

$$Q := \frac{2 \cdot \pi \cdot fr \cdot L}{R} \qquad Q = 125.664$$

SWRref := 2.618

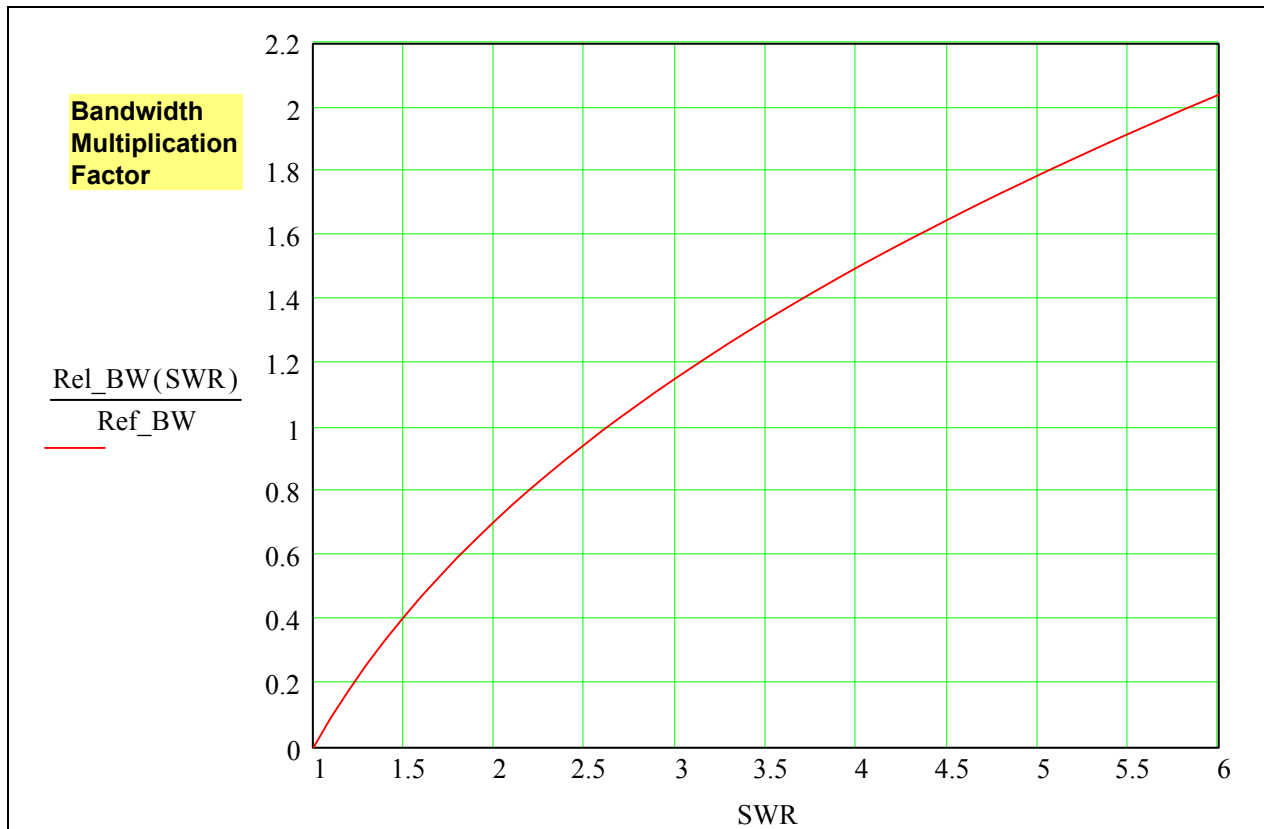
The reference BW is the BW at SWR = 2.618

Ref_BW := Rel_BW(SWRref) Ref_BW = 7.9576 × 10⁻³

$$\frac{1}{\text{Ref_BW}} = Q$$

Relative BW / Reference BW vs SWR

SWR := 1, 1.1..6



For SWR = 2, the BW is 0.707 times the BW at the reference SWR of 2.618

For SWR = 5.83, the BW is 2.0 times the BW at the reference SWR of 2.618

$$\frac{\text{Rel_BW}(2)}{\text{Ref_BW}} = 0.7071$$

$$\frac{\text{Rel_BW}(5.83)}{\text{Ref_BW}} = 2$$

$$\frac{\text{Rel_BW}(4)}{\text{Ref_BW}} = 1.5$$

Pick the SWR value that you like !

For the **series R L C** circuit, a random length of low loss 50 ohm coax cable may also be connected between the RLC and the SWR measuring instrument.
With a coax line, the impedance will differ from 50 +/- j 50 ohms at the SWR = 2.618 points, leaving the SWR bandwidth unchanged.